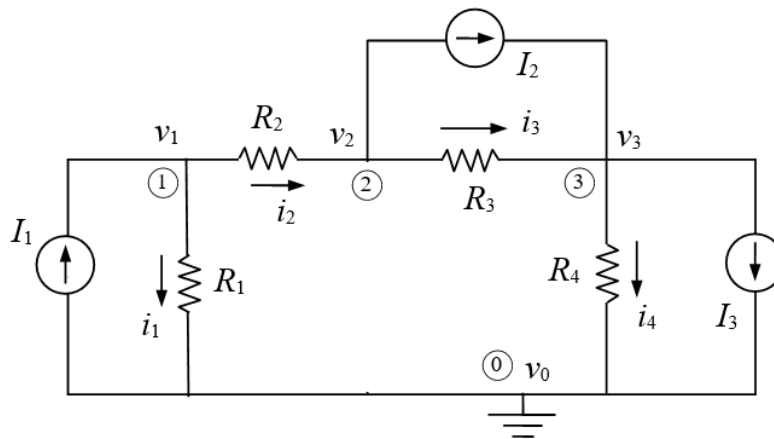


Chap 2 Resistive Circuits

2.1 Node-Voltage Analysis

2.1.1 Independent current sources

- Resistive circuit with only independent current sources I_1 , I_2 and I_3



Reference node: ①, Reference voltage: $v_0 = 0$

Nodes: ①, ②, ③, Node voltages referring to v_0 : v_1 , v_2 , v_3

Unknown voltages to be solved: v_1 , v_2 , v_3

- Node voltage equations

$$\text{KCL}①: \quad i_1 + i_2 - I_1 = \frac{v_1}{R_1} + \frac{v_1 - v_2}{R_2} - I_1 = 0$$

$$\text{KCL}②: \quad -i_2 + i_3 + I_2 = \frac{v_2 - v_1}{R_2} + \frac{v_2 - v_3}{R_3} + I_2 = 0$$

$$\text{KCL}③: \quad -i_3 + i_4 - I_2 + I_3 = \frac{v_3 - v_2}{R_3} + \frac{v_3}{R_4} - I_2 + I_3 = 0$$

- In terms of conductances,

$$\text{KCL}①: \quad (G_1 + G_2)v_1 - G_2 v_2 = I_1$$

$$\text{KCL}②: \quad -G_2 v_1 + (G_2 + G_3)v_2 - G_3 v_3 = -I_2$$

$$\text{KCL}③: \quad -G_3 v_2 + (G_3 + G_4)v_3 = I_2 - I_3$$

- In matrix form,

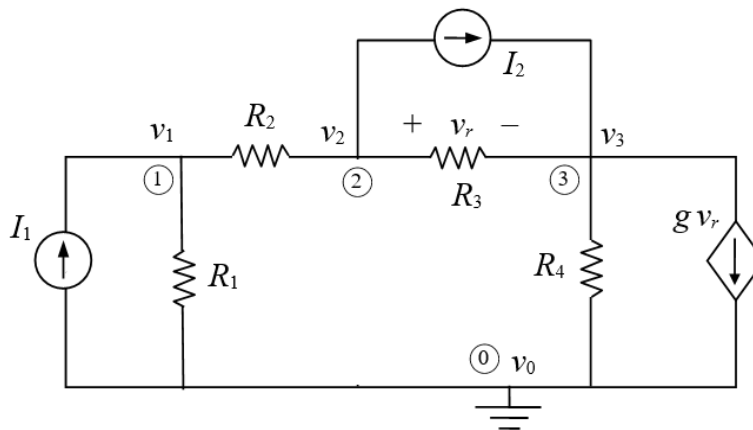
$$(2.1.1-1) \quad \underbrace{\begin{bmatrix} G_1+G_2 & -G_2 & 0 \\ -G_2 & G_2+G_3 & -G_3 \\ 0 & -G_3 & G_3+G_4 \end{bmatrix}}_{\mathbf{A}} \cdot \underbrace{\begin{bmatrix} v_1 \\ v_2 \\ v_3 \end{bmatrix}}_{\mathbf{v}} = \underbrace{\begin{bmatrix} 1 & 0 & 0 \\ 0 & -1 & 0 \\ 0 & 1 & -1 \end{bmatrix}}_{\mathbf{B}} \cdot \underbrace{\begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix}}_{\mathbf{u}}$$

$$\Rightarrow \mathbf{v} = [v_1 \ v_2 \ v_3]^T = \mathbf{A}^{-1} \mathbf{B} \mathbf{u}$$

- \mathbf{A} : symmetric positive-definite, \mathbf{A}^{-1} exists
- All the component's voltages and currents can be obtained in terms of the node voltages. For example, $i_1 = \frac{v_1}{R_1}$, $i_2 = \frac{v_1 - v_2}{R_2}$, $i_3 = \frac{v_2 - v_3}{R_3}$, $i_4 = \frac{v_3}{R_4}$.

2.1.2 Dependent current sources

- Resistive circuit with only current sources including at least one dependent current source such as $g v_r$, where $v_r = v_2 - v_3$.



- Node voltage equations

$$\text{KCL} \textcircled{1}: \quad \frac{v_1}{R_1} + \frac{v_1 - v_2}{R_2} - I_1 = 0$$

$$\text{KCL} \textcircled{2}: \quad \frac{v_2 - v_1}{R_2} + \frac{v_2 - v_3}{R_3} + I_2 = 0$$

$$\text{KCL} \textcircled{3}: \quad -i_3 + i_4 - I_2 + g v_r = \frac{v_3 - v_2}{R_3} + \frac{v_3}{R_4} - I_2 + g(v_2 - v_3) = 0$$

- In terms of conductances,

$$\text{KCL①: } (G_1 + G_2)v_1 - G_2 v_2 = I_1$$

$$\text{KCL②: } -G_2 v_1 + (G_2 + G_3)v_2 - G_3 v_3 = -I_2$$

$$\text{KCL③: } (-G_3 + g)v_2 + (G_3 + G_4 - g)v_3 = I_2$$

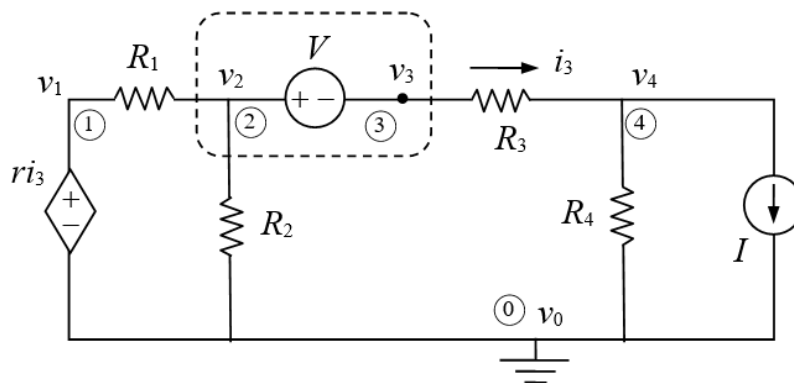
- In matrix form,

$$(2.1.2-1) \quad \underbrace{\begin{bmatrix} G_1 + G_2 & -G_2 & 0 \\ -G_2 & G_2 + G_3 & -G_3 \\ 0 & -G_3 + g & G_3 + G_4 - g \end{bmatrix}}_{\mathbf{A}} \cdot \underbrace{\begin{bmatrix} v_1 \\ v_2 \\ v_3 \end{bmatrix}}_{\mathbf{v}} = \underbrace{\begin{bmatrix} 1 & 0 \\ 0 & -1 \\ 0 & 1 \end{bmatrix}}_{\mathbf{B}} \cdot \underbrace{\begin{bmatrix} I_1 \\ I_2 \end{bmatrix}}_{\mathbf{u}}$$

- If $g \neq G_3 + G_4 + G_3 G_4 \frac{G_1 + G_2}{G_1 G_2}$, then \mathbf{A} is invertible and $\mathbf{v} = \mathbf{A}^{-1} \mathbf{B} \mathbf{u}$.
- All the component's voltages and currents in the resistive circuit can be expressed in terms of the node voltages.

2.1.3 Voltage sources

- Resistive circuit including at least one voltage source such as independent voltage source V and dependent voltage source ri_3 , where $i_3 = \frac{v_3 - v_4}{R_3}$.
- Supernode '②③'.



- Node voltage equations

$$\text{KVL①: } v_1 = ri_3 = r \left(\frac{v_3 - v_4}{R_3} \right)$$

$$\text{KCL}②③ : \frac{v_2 - v_1}{R_1} + \frac{v_2}{R_2} + \frac{v_3 - v_4}{R_3} = 0$$

$$\text{KVL}②③ : v_2 - v_3 = V$$

$$\text{KCL}④ : \frac{v_4 - v_3}{R_3} + \frac{v_4}{R_4} + I = 0$$

- In terms of conductances,

$$\text{KVL}① : v_1 - rG_3v_3 + rG_3v_4 = 0$$

$$\text{KCL}②③ : -G_1v_1 + (G_1 + G_2)v_2 + G_3v_3 - G_3v_4 = 0$$

$$\text{KVL}②③ : v_2 - v_3 = V$$

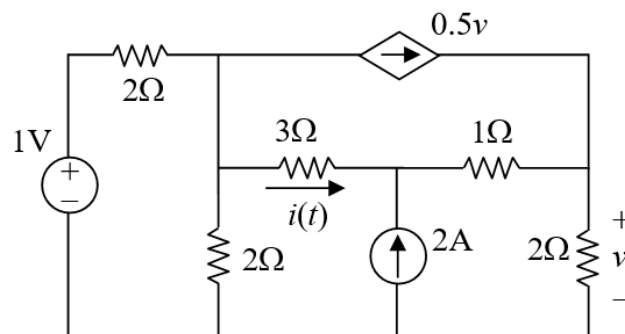
$$\text{KCL}④ : -G_3v_3 + (G_3 + G_4)v_4 = -I$$

- In matrix form,

$$(2.1.3-1) \quad \underbrace{\begin{bmatrix} 1 & 0 & -rG_3 & rG_3 \\ -G_1 & G_1 + G_2 & G_3 & -G_3 \\ 0 & 1 & -1 & 0 \\ 0 & 0 & -G_3 & G_3 + G_4 \end{bmatrix}}_{\mathbf{A}} \cdot \underbrace{\begin{bmatrix} v_1 \\ v_2 \\ v_3 \\ v_4 \end{bmatrix}}_{\mathbf{v}} = \underbrace{\begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 1 & 0 \\ 0 & -1 \end{bmatrix}}_{\mathbf{B}} \cdot \underbrace{\begin{bmatrix} V \\ I \end{bmatrix}}_{\mathbf{u}}$$

- If $r \neq \frac{1}{G_1} + \left(\frac{1}{G_3} + \frac{1}{G_4} \right) \left(1 + \frac{G_2}{G_1} \right)$, then \mathbf{A} is invertible and $\mathbf{v} = \mathbf{A}^{-1}\mathbf{B}\mathbf{u}$.
- All the component's voltages and currents in the resistive circuit can be expressed in terms of the node voltages.

Example: Based on node voltage analysis, determine $i(t)$.

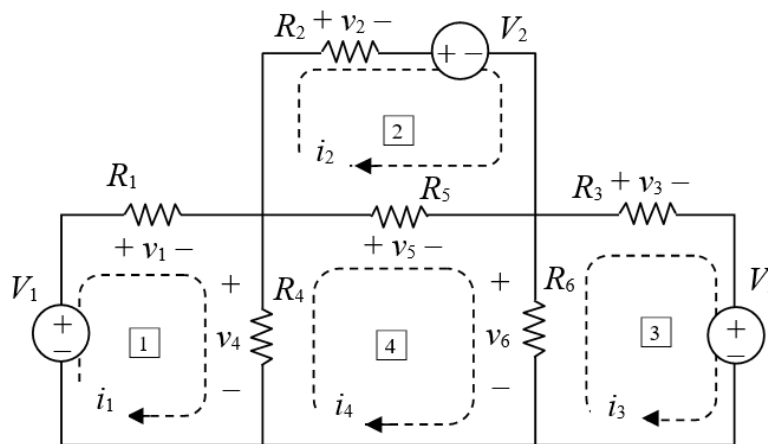


2.2 Mesh-Current Analysis

- Unlike the node-voltage analysis, the mesh-current analysis is only suitable for planar circuit.

2.2.1 Independent voltage sources

- Resistive circuit with only independent voltage sources V_1 , V_2 and V_3



Meshes: [1], [2], [3], [4], Mesh currents: i_1 , i_2 , i_3 , i_4

Unknown currents to be solved: i_1 , i_2 , i_3 , i_4

- Mesh current equations

$$\text{KVL[1]: } v_1 + v_4 - V_1 = R_1 i_1 + R_4 (i_1 - i_4) - V_1 = 0$$

$$\text{KVL[2]: } v_2 + V_2 - v_5 = R_2 i_2 + V_2 - R_5 (i_4 - i_2) = 0$$

$$\text{KVL[3]: } v_3 + V_3 - v_6 = R_3 i_3 + V_3 - R_6 (i_4 - i_3) = 0$$

$$\text{KVL[4]: } v_5 + v_6 - v_4 = R_5 (i_4 - i_2) + R_6 (i_4 - i_3) - R_4 (i_1 - i_4) = 0$$

- In terms of resistance,

$$\text{KVL[1]: } (R_1 + R_4) i_1 - R_4 i_4 = V_1$$

$$\text{KVL[2]: } (R_2 + R_5) i_2 - R_5 i_4 = -V_2$$

$$\text{KVL[3]: } (R_3 + R_6) i_3 - R_6 i_4 = -V_3$$

$$\text{KVL[4]: } -R_4 i_1 - R_5 i_2 - R_6 i_3 + (R_4 + R_5 + R_6) i_4 = 0$$

- In matrix form,

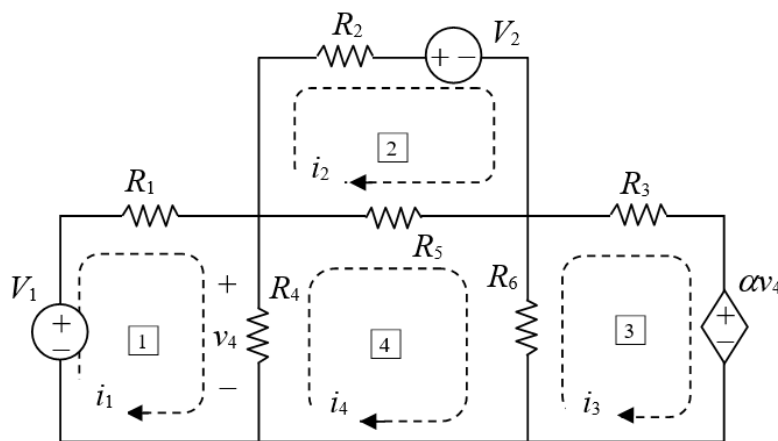
$$(2.2.1-1) \quad \underbrace{\begin{bmatrix} R_1 + R_4 & 0 & 0 & -R_4 \\ 0 & R_2 + R_5 & 0 & -R_5 \\ 0 & 0 & R_3 + R_6 & -R_6 \\ -R_4 & -R_5 & -R_6 & R_4 + R_5 + R_6 \end{bmatrix}}_{\mathbf{A}} \cdot \underbrace{\begin{bmatrix} i_1 \\ i_2 \\ i_3 \\ i_4 \end{bmatrix}}_{\mathbf{i}} = \underbrace{\begin{bmatrix} 1 & 0 & 0 \\ 0 & -1 & 0 \\ 0 & 0 & -1 \\ 0 & 0 & 0 \end{bmatrix}}_{\mathbf{B}} \cdot \underbrace{\begin{bmatrix} V_1 \\ V_2 \\ V_3 \end{bmatrix}}_{\mathbf{u}}$$

$$\Rightarrow \mathbf{i} = [i_1 \ i_2 \ i_3 \ i_4]^T = \mathbf{A}^{-1} \mathbf{B} \mathbf{u}$$

- \mathbf{A} is symmetric positive-definite, \mathbf{A}^{-1} exists.
- All the component's voltages and currents can be obtained in terms of the mesh currents. For example, $v_1 = R_1 i_1$, $v_3 = R_3 i_3$, $v_5 = R_5 (i_4 - i_2)$, $v_6 = R_6 (i_4 - i_3)$.

2.2.2 Dependent voltage sources

- Resistive circuit with only voltage sources including at least one dependent voltage source, such as αv_4 , where $v_4 = R_4 (i_1 - i_4)$.



- Based on KVL, mesh current equations are expressed as

$$\text{KVL[1]: } R_1 i_1 + R_4 (i_1 - i_4) - V_1 = 0$$

$$\text{KVL[2]: } R_2 i_2 + V_2 + R_5 (i_2 - i_4) = 0$$

$$\text{KVL[3]: } R_3 i_3 + \alpha v_4 + R_6 (i_3 - i_4) = R_3 i_3 + \alpha R_4 (i_1 - i_4) + R_6 (i_3 - i_4) = 0$$

$$\text{KVL[4]: } R_5 (i_4 - i_2) + R_6 (i_4 - i_3) + R_4 (i_4 - i_1) = 0$$

- In terms of resistance,

$$\text{KVL[1]: } (R_1 + R_4)i_1 - R_4i_4 = V_1$$

$$\text{KVL[2]: } (R_2 + R_5)i_2 - R_5i_4 = -V_2$$

$$\text{KVL[3]: } \alpha R_4i_1 + (R_3 + R_6)i_3 - (\alpha R_4 + R_6)i_4 = 0$$

$$\text{KVL[4]: } -R_4i_1 - R_5i_2 - R_6i_3 + (R_4 + R_5 + R_6)i_4 = 0$$

- In matrix form,

$$(2.2.2-1) \quad \underbrace{\begin{bmatrix} R_1 + R_4 & 0 & 0 & -R_4 \\ 0 & R_2 + R_5 & 0 & -R_5 \\ \alpha R_4 & 0 & R_3 + R_6 & -\alpha R_4 - R_6 \\ -R_4 & -R_5 & -R_6 & R_4 + R_5 + R_6 \end{bmatrix}}_{\mathbf{A}} \cdot \underbrace{\begin{bmatrix} i_1 \\ i_2 \\ i_3 \\ i_4 \end{bmatrix}}_{\mathbf{i}} = \underbrace{\begin{bmatrix} 1 & 0 \\ 0 & -1 \\ 0 & 0 \\ 0 & 0 \end{bmatrix}}_{\mathbf{B}} \cdot \underbrace{\begin{bmatrix} V_1 \\ V_2 \end{bmatrix}}_{\mathbf{u}}$$

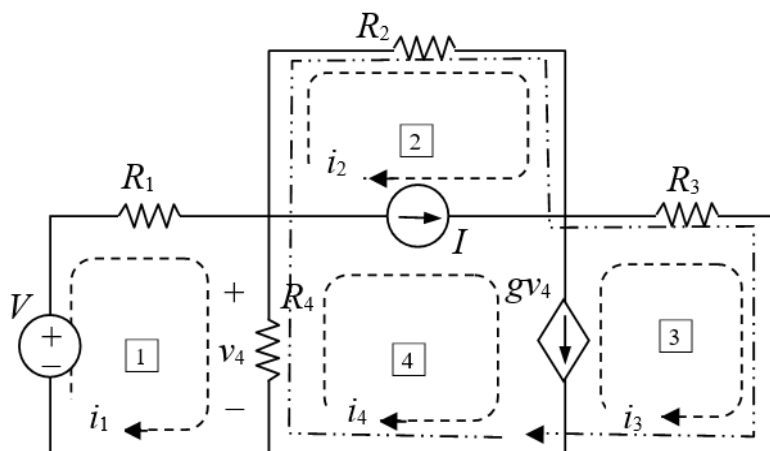
- If $\alpha \neq 1 + \frac{R_3}{R_6} + R_3 \left(\frac{1}{R_1} + \frac{1}{R_4} \right) + \frac{R_2 R_5}{R_2 + R_5} \left(1 + \frac{R_3}{R_6} \right) \left(\frac{1}{R_1} + \frac{1}{R_4} \right)$, then \mathbf{A}^{-1} exists and

$$(2.2.2-2) \quad \mathbf{i} = \mathbf{A}^{-1} \mathbf{B} \mathbf{u}$$

- All the component's voltages and currents in the resistive circuit can be obtained in terms of the mesh currents.

2.2.3 Current sources

- Resistive circuit including at least one current source such as independent current source I and dependent current source gv_4 , where $v_4 = R_4(i_1 - i_4)$.



- Supermesh '234'.

- Mesh current equations

$$\text{KVL}\boxed{1}: \quad R_1 i_1 + R_4 (i_1 - i_4) - V = 0$$

$$\text{KVL}\boxed{2}\boxed{3}\boxed{4}: \quad R_2 i_2 + R_3 i_3 + R_4 (i_4 - i_1) = 0$$

$$\text{KCL}\boxed{2}\boxed{3}\boxed{4}: \quad i_3 - i_4 = -g v_4 = -g R_4 (i_1 - i_4)$$

$$\text{KCL}\boxed{2}\boxed{3}\boxed{4}: \quad i_4 - i_2 = I$$

- In terms of resistance,

$$\text{KVL}\boxed{1}: \quad (R_1 + R_4) i_1 - R_4 i_4 = V$$

$$\text{KVL}\boxed{2}\boxed{3}\boxed{4}: \quad -R_4 i_1 + R_2 i_2 + R_3 i_3 + R_4 i_4 = 0$$

$$\text{KCL}\boxed{2}\boxed{3}\boxed{4}: \quad g R_4 i_1 + i_3 - (1 + g R_4) i_4 = 0$$

$$\text{KCL}\boxed{2}\boxed{3}\boxed{4}: \quad -i_2 + i_4 = I$$

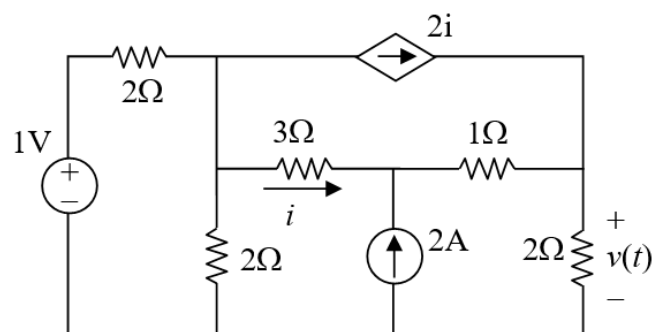
- In matrix form,

$$(2.2.3-1) \quad \underbrace{\begin{bmatrix} R_1 + R_4 & 0 & 0 & -R_4 \\ -R_4 & R_2 & R_3 & R_4 \\ g R_4 & 0 & 1 & -1 - g R_4 \\ 0 & -1 & 0 & 1 \end{bmatrix}}_{\mathbf{A}} \cdot \underbrace{\begin{bmatrix} i_1 \\ i_2 \\ i_3 \\ i_4 \end{bmatrix}}_{\mathbf{i}} = \underbrace{\begin{bmatrix} 1 & 0 \\ 0 & 0 \\ 0 & 0 \\ 0 & 1 \end{bmatrix}}_{\mathbf{B}} \cdot \underbrace{\begin{bmatrix} V \\ I \end{bmatrix}}_{\mathbf{u}}$$

Since $g \neq -\left(\frac{1}{R_1} + \frac{1}{R_4}\right)\left(1 + \frac{R_2}{R_3}\right) - \frac{1}{R_3}$, \mathbf{A} is invertible and $\mathbf{i} = \mathbf{A}^{-1} \mathbf{B} \mathbf{u}$.

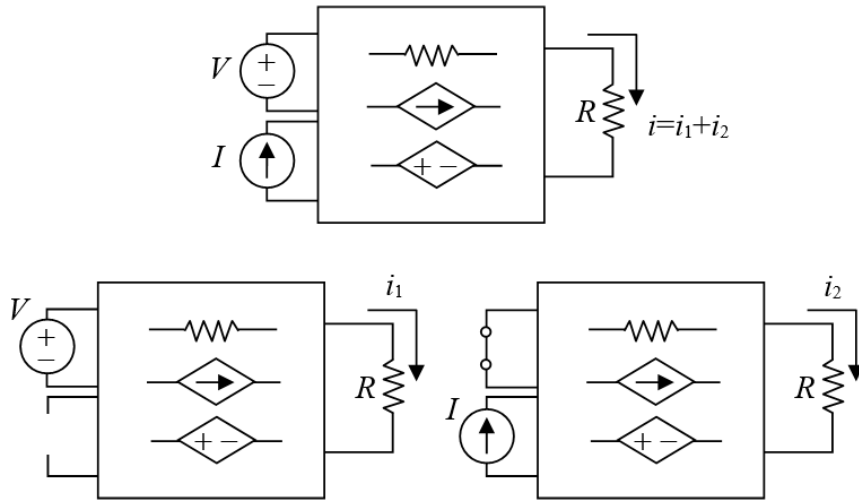
- All the component's voltages and currents in the resistive circuit can be expressed in terms of the mesh currents.

Example: Based on mesh current analysis, determine $v(t)$.

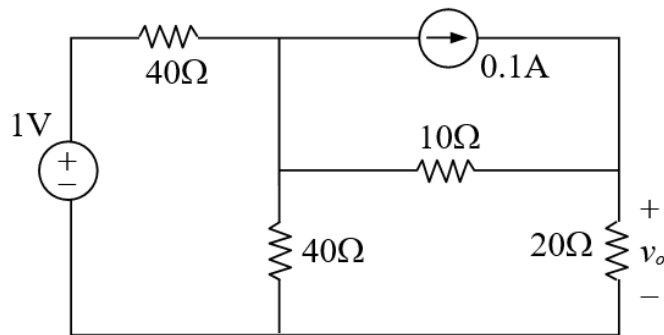


2.3 Principle of Superposition

- Resistive circuit with independent voltage source V and current source I

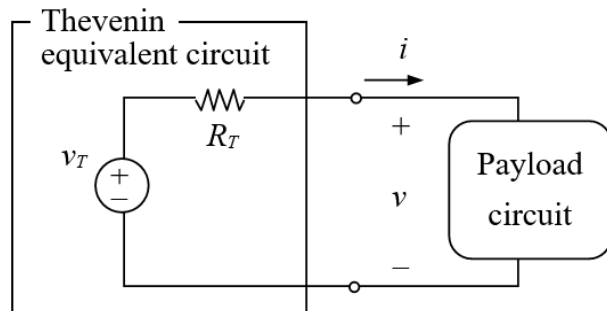
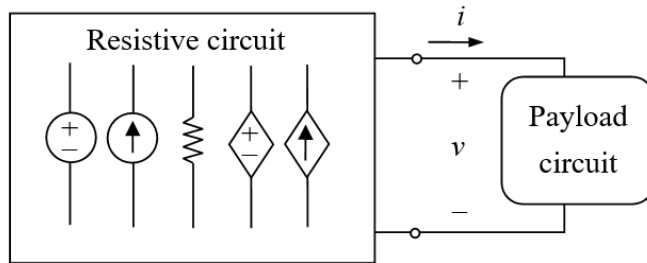


Example: Determine v_o .



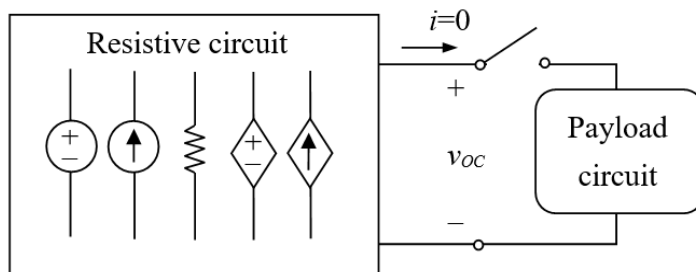
2.4 Thevenin Equivalent Circuits

- Resistive circuit with Independent sources



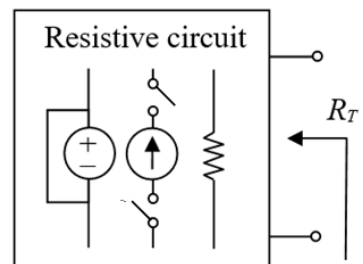
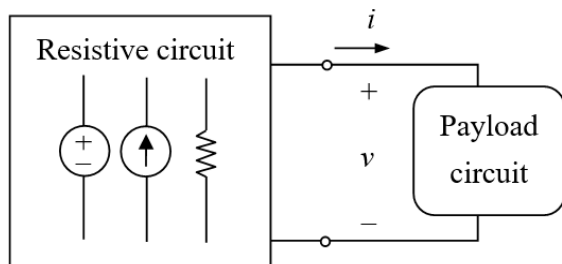
$$(2.4-1) \quad v_T = v + R_T i$$

- Thevenin equivalent voltage $v_T = v_{OC}$

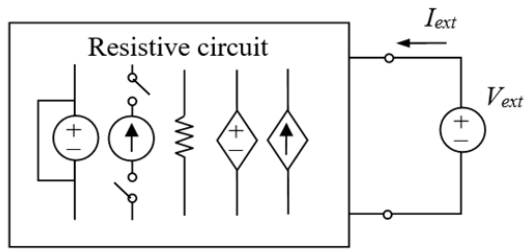


- Thevenin equivalent resistance R_T

[A] Without dependent sources

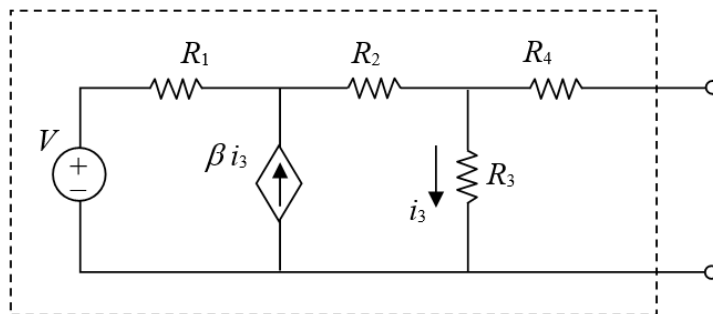
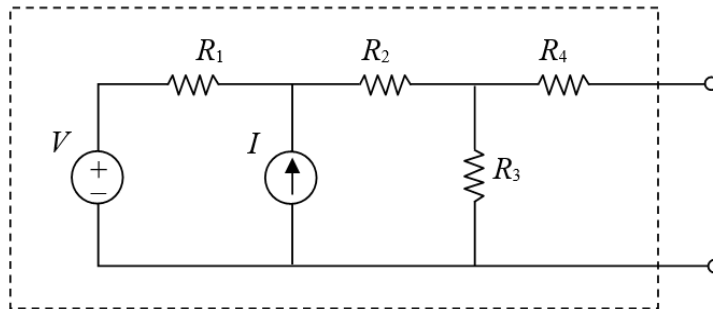
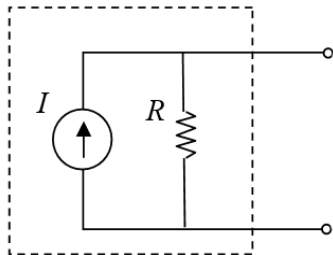


[B] With dependent sources



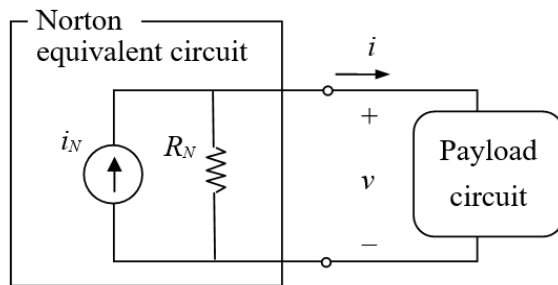
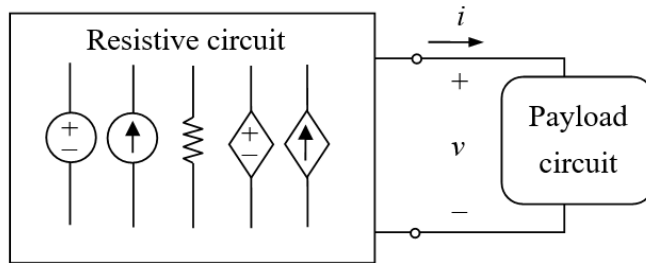
$$(2.4-2) \quad R_T = \frac{V_{ext}}{I_{ext}}$$

Example: Determine the Thevenin equivalent circuit.



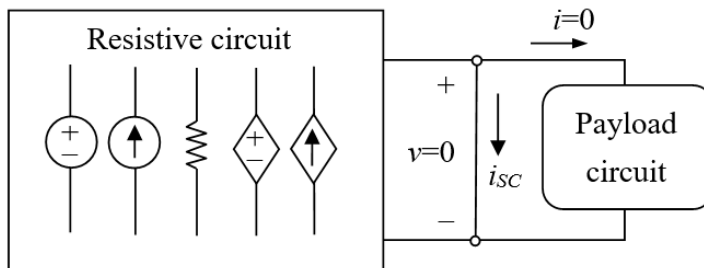
2.5 Norton Equivalent Circuits

- Resistive circuit with independent sources



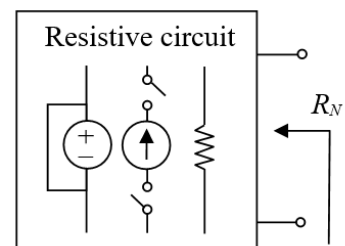
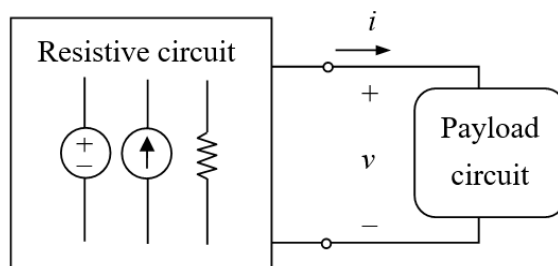
$$(2.5-1) \quad i_N = i + \frac{v}{R_N}$$

- Norton equivalent current $i_N = i_{SC}$

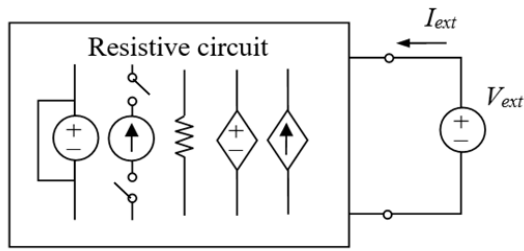


- Norton equivalent resistance R_N

[A] Without dependent sources

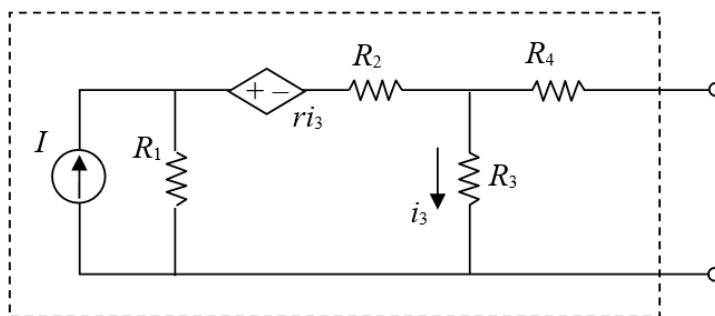
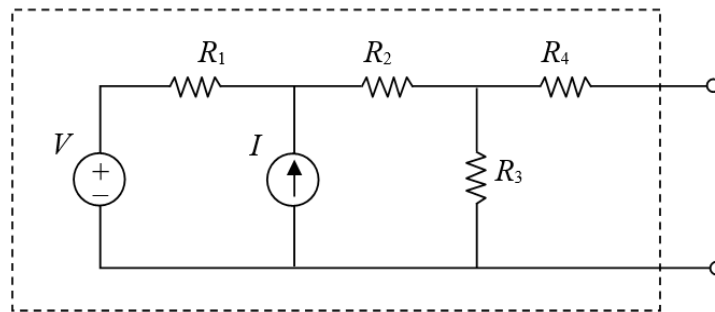
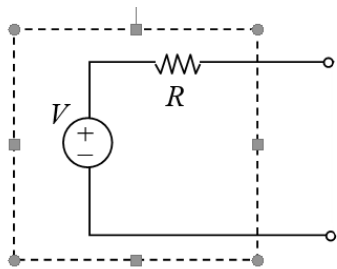


[B] With dependent sources



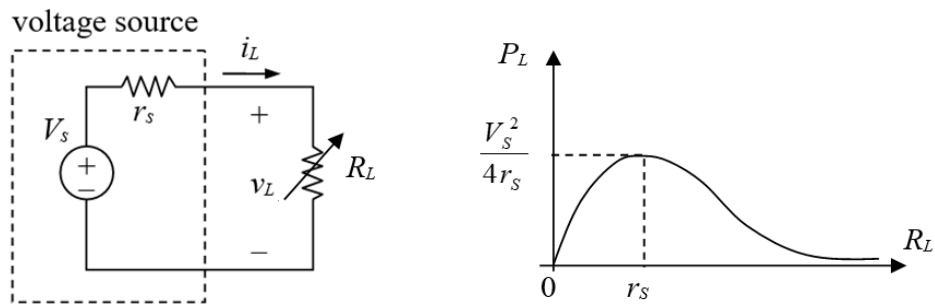
$$(2.5-2) \quad R_N = \frac{V_{ext}}{I_{ext}}$$

Example: Determine the Norton equivalent circuit.



2.6 Maximum Power Transfer Theorem

- The power transferred from voltage source to adjustable R_L



$$(2.6-1) \quad i_L = \frac{V_s}{r_s + R_L}, \quad v_L = R_L i_L = \frac{R_L V_s(t)}{r_s + R_L}$$

The power provided by voltage source

$$(2.6-2) \quad P_s = V_s i_L = \frac{V_s^2}{r_s + R_L}$$

The power transferred to R_L

$$(2.6-3) \quad P_L = i_L v_L = \frac{R_L V_s^2(t)}{(r_s + R_L)^2} = \frac{R_L}{r_s + R_L} P_s$$

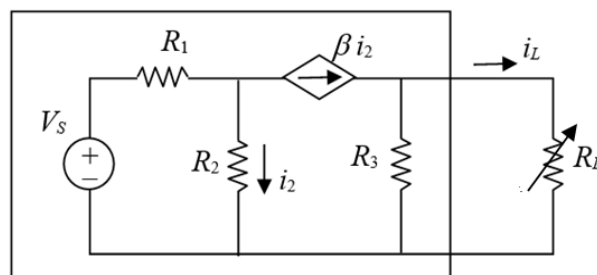
The derivatives of P_L with respect to R_L are

$$\frac{dP_L}{dR_L} = \frac{(r_s - R_L)V_s^2(t)}{(r_s + R_L)^3} \quad \text{and} \quad \frac{d^2P_L}{dR_L^2} = \frac{-2(2r_s - R_L)V_s^2(t)}{(r_s + R_L)^4}$$

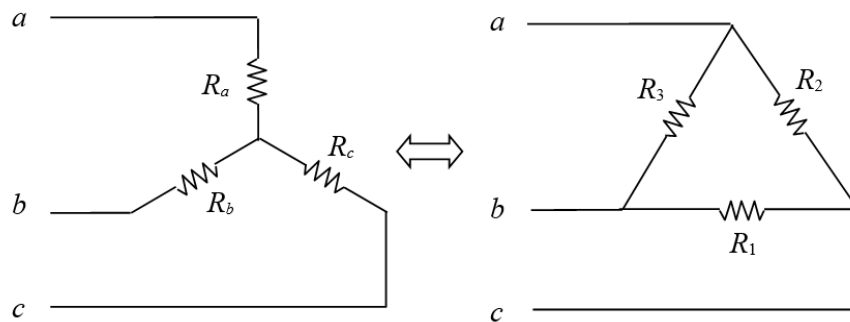
Since $\left. \frac{dP_L}{dR_L} \right|_{R_L=r_s} = 0$ and $\left. \frac{d^2P_L}{dR_L^2} \right|_{R_L=r_s} < 0$, the maximum transferred power is

$$(2.6-4) \quad P_{L,max} = P_L|_{R_L=r_s} = \frac{V_s^2(t)}{4r_s} = \frac{1}{2} P_s|_{R_L=r_s}$$

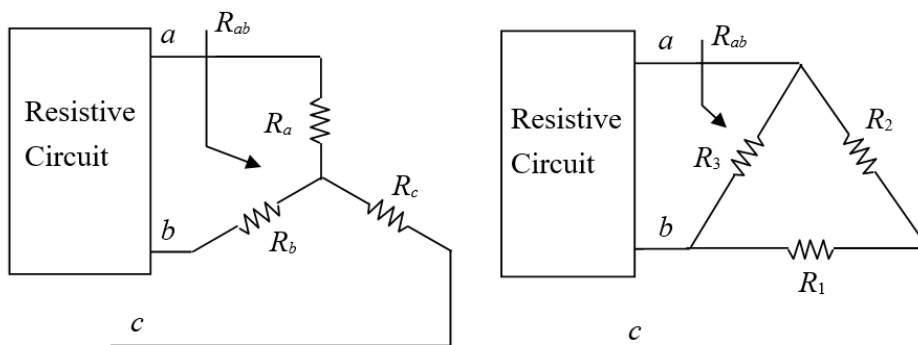
Example: Determine the maximum transferred power to adjustable R_L .



2.7 Y- Δ Resistive Circuit Transformation



- Resistance R_{ab} with terminal c open



$$(2.7-1) \quad R_a + R_b = R_3 \parallel (R_1 + R_2) = \frac{R_3(R_1 + R_2)}{R_1 + R_2 + R_3} = \frac{R_2 R_3 + R_3 R_1}{R_1 + R_2 + R_3}$$

- The relation between Y-type and Δ -type resistive circuits

$$(2.7-2) \quad \begin{cases} R_a + R_b = \frac{R_3(R_1 + R_2)}{R_1 + R_2 + R_3} \\ R_b + R_c = \frac{R_1(R_2 + R_3)}{R_1 + R_2 + R_3} \\ R_c + R_a = \frac{R_2(R_3 + R_1)}{R_1 + R_2 + R_3} \end{cases} \Rightarrow R_a + R_b + R_c = \frac{R_1 R_2 + R_2 R_3 + R_3 R_1}{R_1 + R_2 + R_3}$$

Hence,

$$(2.7-3) \quad R_a = \frac{R_2 R_3}{R_1 + R_2 + R_3}, \quad R_b = \frac{R_3 R_1}{R_1 + R_2 + R_3}, \quad R_c = \frac{R_1 R_2}{R_1 + R_2 + R_3}$$

which results in

$$(2.7-4) \quad R_1 R_a = R_2 R_b = R_3 R_c = \frac{R_1 R_2 R_3}{R_1 + R_2 + R_3}$$

$$(2.7-5) \quad \begin{cases} R_a R_b = \frac{R_3 (R_1 R_2 R_3)}{(R_1 + R_2 + R_3)^2} \\ R_b R_c = \frac{R_1 (R_1 R_2 R_3)}{(R_1 + R_2 + R_3)^2} \\ R_c R_a = \frac{R_2 (R_1 R_2 R_3)}{(R_1 + R_2 + R_3)^2} \end{cases} \Rightarrow R_a R_b + R_b R_c + R_c R_a = \frac{R_1 R_2 R_3}{R_1 + R_2 + R_3}$$

From (2.7-4) and (2.7-5), we have

$$(2.7-6) \quad R_1 R_a = R_2 R_b = R_3 R_c = R_a R_b + R_b R_c + R_c R_a$$

which leads to

$$(2.7-7) \quad R_1 = \frac{R_a R_b + R_b R_c + R_c R_a}{R_a}$$

$$(2.7-8) \quad R_2 = \frac{R_a R_b + R_b R_c + R_c R_a}{R_b}$$

$$(2.7-9) \quad R_3 = \frac{R_a R_b + R_b R_c + R_c R_a}{R_c}$$

$$\underbrace{\frac{R_1 R_2 R_3}{R_1 + R_2 + R_3}}_{\text{from } \Delta \text{ to Y}} = \overbrace{R_1 R_a = R_2 R_b = R_3 R_c = R_a R_b + R_b R_c + R_c R_a}^{\text{from Y to } \Delta}$$

Example: Determine $v(t)$.

